# 16-Non-Ideal Transformers 

Text 11.5
ECEGR 3500
Electrical Energy Systems
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## Magnetizing Reactance

- Non-ideal transformers do not have near infinite permeability

$$
\mathfrak{R}=\frac{\ell}{\mu \mathrm{A}} \neq 0 \quad \mathfrak{I}=\mathrm{N}_{1} \mathbf{I}_{1}-\mathrm{N}_{2} \mathbf{I}_{2}=\mathfrak{R} \Phi_{\mathrm{m}} \neq 0
$$

- Add shunt magnetizing reactance $\left(\mathrm{X}_{0}\right)$ to ideal transformer model $I_{i}$



## Magnetizing Reactance

without load


$$
\begin{gathered}
\mathrm{N}_{1} \boldsymbol{I}_{2}^{\prime}=\mathfrak{R} \boldsymbol{\Phi}_{\mathrm{m}}+\mathrm{N}_{2} \boldsymbol{I}_{2} \\
\quad \text { (vector sum) }
\end{gathered}
$$



## Exercise

- A transformer has 450 turns on the primary and 50 turns on the secondary. The primary voltage is 6000 V . If the magnetizing reactance is $j 500 \Omega$ compute:
- The no-load primary current and real power loss of the transformer
- The primary current if a load impedance of $\boldsymbol{Z}=10+\mathrm{jl5}$ is applied to the secondary.


## Exercise

$$
\boldsymbol{I}_{1}=\boldsymbol{I}_{0}+\boldsymbol{I}_{2}^{\prime}=\frac{\boldsymbol{V}_{1}}{500 \angle 90^{\circ}}+0=12 \angle-90^{\circ} \mathrm{A}
$$

$$
\begin{aligned}
& \mathrm{P}_{\text {in }}=\operatorname{Re}\left\{\boldsymbol{V}_{1} \boldsymbol{I}_{1}^{*}\right\}=0 \mathrm{~W} \\
& \mathrm{P}_{\text {out }}=\operatorname{Re}\left\{\boldsymbol{V}_{2} \boldsymbol{I}_{2}^{*}\right\}=0 \mathrm{~W} \\
& \mathrm{P}_{\text {Loss }}=\mathrm{P}_{\text {in }}-\mathrm{P}_{\text {out }}=0 \mathrm{~W}
\end{aligned}
$$



## Exercise

$$
\begin{aligned}
& \boldsymbol{I}_{0}=\frac{\boldsymbol{V}_{1}}{500 \angle 90^{\circ}}=12 \angle-90^{\circ} \mathrm{A} \\
& \boldsymbol{E}_{1}=\boldsymbol{V}_{1}=\frac{\mathrm{N}_{1}}{\mathrm{~N}_{2}} \boldsymbol{E}_{2} \\
& \boldsymbol{E}_{2}=667 \angle 0^{\circ} \mathrm{V}
\end{aligned}
$$

$$
\boldsymbol{I}_{2}=\frac{\boldsymbol{E}_{2}}{10+j 15}=36.98 \angle-56.3^{\circ} \mathrm{A}
$$

$$
\boldsymbol{I}_{1}=\boldsymbol{I}_{0}+\boldsymbol{I}_{2}^{\prime}=\boldsymbol{I}_{0}+\frac{50}{450} \boldsymbol{I}_{2}=15.59 \angle-81.6^{\circ} \mathrm{A}
$$



## Core Resistance

- Non-ideal transformers have eddy current loss
- real power loss
- occurs even with no secondary load
- Model as shunt resistance
- $\mathrm{R}_{0} \gg \mathrm{X}_{0}$


Note: xfmr are designed to have large $\mathrm{X}_{0}$, $\mathrm{R}_{0}$ values

## Leakage Flux

- Non-ideal transformers have leakage flux
- Leakage flux: flux in primary(secondary) coil that is not linked to secondary (primary) coil

$$
\begin{aligned}
& \boldsymbol{\lambda}_{1}=\boldsymbol{\lambda}_{11}+\mathrm{N}_{1} \boldsymbol{\Phi}_{\mathrm{m}} \\
& \boldsymbol{\lambda}_{2}=-\boldsymbol{\lambda}_{12}+\mathrm{N}_{2} \boldsymbol{\Phi}_{\mathrm{m}} \\
& \boldsymbol{V}_{1}=\frac{\mathrm{d} \boldsymbol{\lambda}_{1}}{\mathrm{~d} t}=L_{11} \frac{\mathrm{~d} \boldsymbol{I}_{1}}{\mathrm{~d} t}+\mathrm{N}_{1} \frac{\mathrm{~d} \boldsymbol{\Phi}_{\mathrm{m}}}{\mathrm{~d} t} \\
& \boldsymbol{V}_{2}=\frac{\mathrm{d} \boldsymbol{\lambda}_{2}}{\mathrm{~d} t}=-L_{12} \frac{\mathrm{~d} \boldsymbol{I}_{2}}{\mathrm{~d} t}+\mathrm{N}_{2} \frac{\mathrm{~d} \boldsymbol{\Phi}_{\mathrm{m}}}{\mathrm{~d} t}
\end{aligned}
$$



## Leakage Flux

- Model as series reactances on primary and secondary
- Xmfrs are generally designed to have low leakage reactance
- $\mathrm{X}_{1} \ll \mathrm{X}_{0}$



## Winding Resistance

- Include winding resistance
- $\mathrm{R}_{1}<X_{1}, R_{2}<X_{2}$

$$
\begin{aligned}
& \boldsymbol{V}_{1}=\mathrm{R}_{1} \boldsymbol{I}_{1}+\frac{\mathrm{d} \boldsymbol{\lambda}_{1}}{\mathrm{~d} t}=\mathrm{R}_{1} \boldsymbol{I}_{1}+L_{1} \frac{\mathrm{~d} \boldsymbol{I}_{1}}{\mathrm{~d} t}+\mathrm{N}_{1} \frac{\mathrm{~d} \boldsymbol{\Phi}_{\mathrm{m}}}{\mathrm{~d} t} \\
& \boldsymbol{v}_{2}=-\mathrm{R}_{2} \boldsymbol{I}_{2}+\frac{\mathrm{d} \boldsymbol{I}_{2}}{\mathrm{~d} t}=-\mathrm{R}_{2} \boldsymbol{I}_{2}-L_{2} \frac{\mathrm{~d} \boldsymbol{I}_{2}}{\mathrm{~d} t}+\mathrm{N}_{2} \frac{\mathrm{~d} \boldsymbol{\Phi}_{\mathrm{m}}}{\mathrm{~d} t}
\end{aligned}
$$



## Winding Resistance

If you were designing a transformer with the shown number of turns, would you rather:
A. use the same gauge wire on the primary and secondary
B. use larger diameter on the primary
C. use larger diameter wire on the secondary


## Winding Resistance

More current is flowing through the secondary, so it requires lower resistance to dissipate the same heat. You should use larger diameter wire.

```
Side with fewer turns (lower voltage, higher current)
    has lower resistance wire
```



## Example

- Let:
- $\mathrm{X}_{\mathrm{o}}=20,000 \Omega$
- $\mathrm{R}_{\mathrm{o}}=40,000 \Omega$
- $\mathrm{R}_{1}=2.56 \Omega$
- $\mathrm{R}_{2}=0.010 \Omega$
- $\mathrm{X}_{1}=3.84 \Omega$
- $\mathrm{X}_{2}=0.015 \Omega$
- $\mathrm{N}_{1}=3200$
- $\mathrm{N}_{2}=200$

Find $\boldsymbol{I}_{1}$, and the input power

- $\mathrm{Z}_{\text {load }}=10 \Omega$
- $\left|\mathbf{V}_{1}\right|=8000 \mathrm{~V}$



## Example

- Let:
- $X_{o}=20,000 \Omega$
- $\mathrm{N}_{1}=3200$
- $\mathrm{R}_{\mathrm{o}}=40,000 \Omega$
- $\mathrm{N}_{2}=200$
- $\mathrm{R}_{1}=2.56 \Omega$
- $\mathrm{Z}_{\text {load }}=10 \Omega$
- $\mathrm{R}_{2}=0.010 \Omega$
- $\left|\mathrm{V}_{1}\right|=8000 \mathrm{~V}$

- $\mathrm{X}_{1}=3.84 \Omega$
- $\mathrm{X}_{2}=0.015 \Omega$
$a=\frac{N_{1}}{N_{2}}=\frac{3200}{200}=16$
$Z_{2}=R_{2}+j X_{2}+Z_{\text {load }}=10.01+j 0.015 \Omega$
$Z_{2}^{\prime}=a^{2} Z_{2}=2563+j 3.84 \Omega$



## Example

- Let:
- $\mathrm{X}_{\mathrm{o}}=20,000 \Omega$
- $\mathrm{N}_{1}=3200$
- $\mathrm{R}_{\mathrm{o}}=40,000 \Omega$
- $\mathrm{N}_{2}=200$
- $\mathrm{R}_{1}=2.56 \Omega$
- $\mathrm{Z}_{\text {load }}=10 \Omega$
- $\mathrm{R}_{2}=0.010 \Omega$
- $\left|\mathrm{V}_{1}\right|=8000 \mathrm{~V}$

- $\mathrm{X}_{1}=3.84 \Omega$
- $\mathrm{X}_{2}=0.015 \Omega$

$$
Z_{e q 1}=R_{1}+j X_{1}+Z_{2}^{\prime}=2565+j 3.855 \Omega
$$



## Example

- Let:
- $X_{o}=20,000 \Omega$
- $\mathrm{N}_{1}=3200$
- $\mathrm{R}_{\mathrm{o}}=40,000 \Omega$
- $\mathrm{N}_{2}=200$
- $\mathrm{R}_{1}=2.56 \Omega$
- $\mathrm{Z}_{\text {load }}=10 \Omega$
- $\mathrm{R}_{2}=0.010 \Omega$
- $\left|\mathrm{V}_{1}\right|=8000 \mathrm{~V}$

- $\mathrm{X}_{1}=3.84 \Omega$
- $\mathrm{X}_{2}=0.015 \Omega$
$Z_{e q}=R_{0}\left\|j X_{0}\right\| Z_{e q 1}$
$\frac{1}{Z_{e q}}=\frac{1}{R_{0}}+\frac{1}{j X_{0}}+\frac{1}{Z_{e q 1}}$
$Z_{e q}=2375+j 289.6 \Omega$


## Example

- Let:
- $\mathrm{X}_{\mathrm{o}}=20,000 \Omega \quad \cdot \mathrm{~N}_{1}=3200$
- $\mathrm{R}_{\mathrm{o}}=40,000 \Omega$
- $\mathrm{N}_{2}=200$
- $\mathrm{R}_{1}=2.56 \Omega$
- $\mathrm{Z}_{\text {load }}=10 \Omega$
- $\mathrm{R}_{2}=0.010 \Omega$
- $\left|\mathrm{V}_{1}\right|=8000 \mathrm{~V}$
- $\mathrm{X}_{1}=3.84 \Omega$
- $\mathrm{X}_{2}=0.015 \Omega$

$$
\begin{aligned}
& \boldsymbol{I}_{1}=\frac{\boldsymbol{V}_{1}}{Z_{e q}}=3.32-j 0.405=3.34 \angle-6.95 \mathrm{~A} \\
& \mathrm{P}_{\text {in }}=\operatorname{Re}\left\{\boldsymbol{V}_{1} \boldsymbol{I}_{1}^{*}\right\}=26,550 \mathrm{~W}
\end{aligned}
$$



## Example

- Now find the output power
- Let:
- $X_{o}=20,000 \Omega$
- $\mathrm{N}_{1}=3200$
- $\mathrm{R}_{\mathrm{o}}=40,000 \Omega$
- $\mathrm{R}_{1}=2.56 \Omega$
- $\mathrm{R}_{2}=0.010 \Omega$
- $\mathrm{X}_{1}=3.84 \Omega$
- $\mathrm{X}_{2}=0.015 \Omega$
- $\mathrm{N}_{2}=200$
- $\mathrm{Z}_{\text {load }}=10 \Omega$
- $\left|\mathbf{V}_{1}\right|=8000 \mathrm{~V}$



## Example

- Now find the output power

$$
\begin{aligned}
& \boldsymbol{I}_{1}=\frac{\boldsymbol{V}_{1}}{Z_{e q}}=3.32-j 0.405=3.34 \angle-6.95 \mathrm{~A} \\
& \boldsymbol{I}_{0}=\frac{\boldsymbol{V}_{1}}{R_{0}}+\frac{\boldsymbol{V}_{1}}{j X_{0}}=0.2-j 0.4 \Omega
\end{aligned}
$$



$$
\begin{aligned}
& \boldsymbol{I}_{2}^{\prime}=\boldsymbol{I}_{1}-\boldsymbol{I}_{0}=3.12-j 0.005=3.12 \angle-0.09^{\circ} \mathrm{A} \\
& \boldsymbol{I}_{2}=\boldsymbol{I}_{2}^{\prime} a=49.9-j 0.075=49.9 \angle-0.09^{\circ} \mathrm{A} \\
& \mathrm{P}_{\text {out }}=\left|\boldsymbol{I}_{2}\right|^{2} \operatorname{Re}\left\{Z_{\text {load }}\right\}=24,900 \mathrm{~W}
\end{aligned}
$$

## Approximate Circuit

- Often desirable to simplify the transformer model
- More accurate than ideal, less accurate than exact
- Voltage drop across $\boldsymbol{Z}_{1}=R_{1}+j X_{1}$ is designed to be small



## Approximate Circuit

- Move $Z_{1}$ to other side of shunt elements
- Next, eliminate the ideal transformer by referring the secondary elements to the primary



## Approximate Circuit

- Move $Z_{1}$ to other side of shunt elements
- Next, eliminate the ideal transformer by referring the secondary elements to the primary



## Approximate Circuit

- Letting:
- $R^{\prime}{ }_{2}=a^{2} R_{2}$
- $X_{2}=\mathrm{a}^{2} X_{2}$
- $Z_{\mathrm{L}}^{\prime}=\mathrm{a}^{2} Z_{\mathrm{L}}$
- $V_{2}=\mathrm{a} V_{2}$



## Approximate Circuit

- Combine series elements

$$
\begin{aligned}
& R_{\mathrm{e} 1}=R_{1}+R_{2}^{\prime} \\
& X_{\mathrm{e} 1}=X_{1}+X_{2}^{\prime}
\end{aligned}
$$

It is also possible to refer to the impedances from the secondary side.


## Approximate Circuit

- Further approximations are possible
- Ignore shunt branch
- Ignore resistances
- Problem statement will indicate which model to use



## Example

- Consider a single-phase xfmr with the following specifications:
- primary turns: 10
- secondary turns: 5
- winding resistance: 0.2 Ohms
- leakage reactance: 0.6 Ohms
- infinite permeability
- If the primary is connected to a 1000 V source and the secondary to a 5 Ohm load, find the power supplied to the load
- Assume the xfmr impedances are referred from the primary and include the secondary impedances


## Example



First calculate the ratio:

$$
a=\frac{\mathrm{N}_{1}}{\mathrm{~N}_{2}}=2
$$

transform the impedance
$\boldsymbol{Z}_{1}=\mathrm{a}^{2} \boldsymbol{Z}_{2}=20 \Omega$

redraw the circuit

## Example



$$
\begin{aligned}
& \boldsymbol{I}_{1}=\frac{1000 \angle 0}{20.2+j 0.6}=49.48 \angle-1.7^{\circ} \mathrm{A} \\
& \mathrm{P}=\left|\boldsymbol{I}_{1}^{2}\right| Z_{\mathrm{L}}^{\prime}=48.97 \mathrm{~kW}
\end{aligned}
$$

## Example

Another approach keeping
the ideal transformer element:

$$
\begin{aligned}
& \boldsymbol{I}_{1}=\frac{1000 \angle 0}{20.2+j 0.6}=49.48 \angle-1.7^{\circ} \mathrm{A} \\
& \boldsymbol{E}_{1}=\boldsymbol{V}_{1}-\boldsymbol{I}_{1}(0.2+\mathrm{j} 0.6)=989.66 \angle-1.7^{\circ} \mathrm{V} \\
& \boldsymbol{V}_{2}=\boldsymbol{E}_{2}=\left(989.66 \angle-1.7^{\circ}\right)\left(\frac{1}{\mathrm{a}}\right)=494.83 \angle-1.7 \mathrm{~V} \\
& \boldsymbol{I}_{2}=\left(49.48 \angle-1.7^{\circ}\right)(2)=98.96 \angle-1.7^{\circ} \mathrm{A} \\
& \mathrm{P}_{2}=\left|\mathbf{V}_{2}\right|\left|\mathbf{I}_{2}\right| \cos (0)=48.97 \mathrm{~kW}
\end{aligned}
$$

## Reading [on your own]

- 11.5.2 Transformer Efficiency
- 11.5.3 Voltage Regulation


## Summary

- Non-ideal xfmrs include: magnetization reactance, leakage reactance, winding resistance and core loss
- Approximations can be made to simplify circuit analysis (series impedances are small, shunt impedances are large)

